

# Data Broadcasting Algorithms on Error-Prone Wireless Channels

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**Abstract.** Broadcasting is an efficient and scalable way of transmitting data over wireless channels to an unlimited number of clients. In this paper the problem of allocating data to multiple channels is studied, assuming flat data scheduling per channel and the presence of unrecoverable channel transmission errors. The objective is that of minimizing the average expected delay experienced by the clients. Two different channel error models are considered: the Bernoulli model and the simplified Gilbert-Elliot one. In the former model, each packet transmission has the same probability to fail and each transmission error is independent from the others. In the latter one, bursts of erroneous or error-free packet transmissions due to wireless fading channels are modeled. For both channel error models, optimal solutions can be found in polynomial time when all data items have unit lengths, while heuristics are presented when data items have non-unit lengths. Extensive simulations, performed on benchmarks whose item popularities follow Zipf distributions, show that good sub-optimal solutions are found.

**Keywords.** Wireless communication, Data broadcasting, Multiple channels, Flat scheduling, Average expected delay, Channel transmission errors, Bernoulli model, Gilbert-Elliot model, Heuristics

## Introduction

In wireless asymmetric communications, *data broadcasting* is an efficient way of simultaneously disseminating data items to a large number of clients [17]. Consider data services on cellular networks, such as stock quotes, weather infos, traffic news, where data are continuously broadcast to clients that may desire them at any instant of time. In this scenario, a server at the base-station repeatedly transmits data items from a given set over wireless channels, while clients passively listen to the shared channels waiting for their desired item. The server has to pursue a data allocation strategy for assigning items to channels and a broadcast schedule for deciding which item has to be transmitted on each channel at any time instant. Therefore, efficient data allocation and broadcast scheduling algorithms have to minimize the client expected delay, that is, the average amount of time spent by a client before receiving the item he needs. Such a delay increases with the size

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of the set of the data items to be transmitted by the server. Indeed, the client has to wait for many unwanted data before receiving his own data. Moreover, the client expected delay may be influenced by transmission errors because items are not always received correctly by the client. Although data are usually encoded using *error correcting codes (ECC)* allowing some recoverable errors to be corrected by the client without affecting the average expected delay, there are several transmission errors which still cannot be corrected using ECC. Such *unrecoverable* errors affect the client expected delay, because the resulting corrupted items have to be discarded and the client must wait until the same item is broadcast again by the server.

Several variants for the problem of data allocation and broadcast scheduling have been proposed in the literature [1,2,3,4,5,6,9,10,11,13,15,16,19,21,22].

The database community usually partitions the data among the channels and then adopts a *flat* broadcast schedule on each channel [5,15,22]. In such a way, the allocation of data to channels becomes critical for reducing the average expected delay, while the flat schedule on each channel merely consists in cyclically broadcasting in an arbitrary fixed order, that is once at a time in a round-robin fashion, the items assigned to the same channel [1]. In order to reduce the average expected delay, a *skewed* data allocation is used where items are partitioned according to their popularities so that the most requested items appear in a channel with shorter period. Assuming that each item transmitted by the server is always received correctly by the client, a solution that minimizes the average expected delay can be found in polynomial time in the case of *unit lengths* [22], that is when all the items have a unit transmission time, whereas the problem becomes computationally intractable for non-unit lengths [5]. In this latter case, several heuristics have been developed in [4,22], which have been tested on some benchmarks where item popularities follow Zipf distributions. Such distributions are used to characterize the popularity of one item among a set of similar data, like a web page in a web site [8].

The data allocation problem has not been investigated by the database community when the wireless channels are subject to transmission errors. In contrast, a wireless environment subject to errors has been considered by the networking community, which however concentrates only on finding broadcast scheduling for a single channel to minimize the average expected delay [6,10,11,19]. Indeed, the networking community assumes all items replicated over all channels, and therefore no data allocation to the channels is needed. Although it is still unknown whether a broadcast schedule on a single channel with minimum average delay can be found in polynomial time or not, almost all the proposed solutions follow the *square root rule (SRR)*, a heuristic which in practice finds near-optimal schedules [3]. The aim of SRR is to produce a broadcast schedule where each data item appears with equally spaced replicas, whose frequency is proportional to the square root of its popularity and inversely proportional to the square root of its length. In particular, the solution proposed by [19] adapts the SRR solution to the case of unrecoverable errors. In such a case, since corrupted items must be discarded worsening the average expected delay, the spacing among replicas has to be properly recomputed.

The present paper considers the data allocation problem under the assumption of flat data schedule per channel [4,5,22], as studied by the database community, but also copes with the presence of unrecoverable erroneous transmissions, as studied in [7,19]. The behavior of wireless channels is described by means of two different error models: the Bernoulli model and the simplified Gilbert-Elliot one [20]. In the former each packet

transmission has the same probability  $q$  to fail and  $1-q$  to succeed, and each transmission error is independent from the others. In contrast, the latter model is able to capture burstiness, that is sequences of erroneous or error-free packet transmissions, and well approximates the error characteristics of certain wireless fading channels [18]. For both channel error models, it is shown that an optimum solution, namely one minimizing the average expected delay, can be found in polynomial time for the data allocation problem when the data items have unit lengths. Instead, sub-optimal solutions found by heuristic algorithms are exhibited for both channel error models and items with non-unit lengths. Extensive simulations show that such heuristics provide good sub-optimal solutions when tested on benchmarks whose items popularities are characterized by Zipf distributions. Moreover, it is proved that optimal solutions can be found in pseudo-polynomial time when there are only two channels, the items have non-unit lengths, and the Bernoulli channel error model is used.

The rest of this paper is so organized. Section 1 first gives notations, definitions as well as the problem statement, and then reviews the basic algorithms known so far in the case of error-free channel transmissions. Sections 2 and 3 consider the Bernoulli and the Gilbert-Elliott channel error model, respectively, and illustrate how the previously reviewed algorithms can be adapted to cope with erroneous transmissions. Experimental evaluations of the algorithms are reported at the end of both Sections 2 and 3. Finally, conclusions are offered in Section 4.

## 1. Error-Free Channels

Consider a set of  $K$  identical error-free channels, and a set  $D = \{d_1, d_2, \dots, d_N\}$  of  $N$  data items. Each item  $d_i$  is characterized by a *popularity*  $p_i$  and a *length*  $z_i$ , with  $1 \leq i \leq N$ . The popularity  $p_i$  represents the demand probability of item  $d_i$ , namely its probability to be requested by the clients, and it does not vary along the time. Clearly,  $\sum_{i=1}^N p_i = 1$ . The length  $z_i$  is an integer number, counting how many packets are required to transmit item  $d_i$  on any channel and it includes the encoding of the item with an error correcting code. For the sake of simplicity, it is assumed that a packet transmission requires one time unit. Each  $d_i$  is assumed to be non preemptive, that is, its transmission cannot be interrupted. When all data lengths are equal to one, i.e.,  $z_i = 1$  for  $1 \leq i \leq N$ , the lengths are called *unit* lengths, otherwise they are said to be *non-unit* lengths. The sum of all the item lengths and the maximum item length are denoted, respectively, by  $Z$  and  $z$ , namely  $Z = \sum_{i=1}^N z_i$  and  $z = \max_{1 \leq i \leq N} z_i$ .

The *expected delay*  $t_i$  is the expected number of packets a client must wait for receiving item  $d_i$ . The *average expected delay (AED)* is the number of packets a client must wait on the average for receiving any item, and is computed as the sum over all items of their expected delay multiplied by their popularity, that is

$$\text{AED} = \sum_{i=1}^N t_i p_i \quad (1)$$

When the items are partitioned into  $K$  groups  $G_1, \dots, G_K$ , where group  $G_k$  collects the data items assigned to channel  $k$ , and a flat schedule is adopted for each channel, that is, the items in  $G_k$  are cyclically broadcast in an arbitrary fixed order, Equation 1 can

be simplified. Indeed, if item  $d_i$  is assigned to channel  $k$ , and assuming that clients can start to listen at any instant of time with the same probability, then  $t_i$  becomes  $\frac{Z_k}{2}$ , where  $Z_k$  is the schedule *period* on channel  $k$ , i.e.,  $Z_k = \sum_{d_i \in G_k} z_i$ . Then, Equation 1 can be rewritten as

$$\text{AED} = \sum_{i=1}^N t_i p_i = \sum_{k=1}^K \sum_{d_i \in G_k} \frac{Z_k}{2} p_i = \sum_{k=1}^K \left( \frac{Z_k}{2} \sum_{d_i \in G_k} p_i \right) = \frac{1}{2} \sum_{k=1}^K Z_k P_k \quad (2)$$

where  $P_k$  denotes the sum of the popularities of the items assigned to channel  $k$ , i.e.,  $P_k = \sum_{d_i \in G_k} p_i$ . Note that, in the unit length case, the period  $Z_k$  coincides with the cardinality of  $G_k$ , which will be denoted by  $N_k$ .

Summarizing, given  $K$  error-free channels, a set  $D$  of  $N$  items, where each data item  $d_i$  comes along with its popularity  $p_i$  and its integer length  $z_i$ , the *Data Allocation Problem* consists in partitioning  $D$  into  $K$  groups  $G_1, \dots, G_K$ , so as to minimize the AED objective function given in Equation 2. Note that, in the special case of unit lengths, the corresponding objective function is derived replacing  $Z_k$  with  $N_k$  in Equation 2.

Almost all the algorithms proposed so far for the data allocation problem on multiple error-free channels are based on dynamic programming. Such algorithms restrict the search for the solutions to the so called *segmentations*, that is, partitions obtained by considering the items ordered by their indices, and by assigning items with consecutive indices to each channel. Formally, a segmentation is a partition of the ordered sequence  $d_1, \dots, d_N$  into  $K$  adjacent segments  $G_1, \dots, G_K$ , each of consecutive items, as follows:

$$\underbrace{d_1, \dots, d_{B_1}}_{G_1}, \underbrace{d_{B_1+1}, \dots, d_{B_2}}_{G_2}, \dots, \underbrace{d_{B_{K-1}+1}, \dots, d_N}_{G_K}$$

A segmentation can be compactly denoted by the  $(K - 1)$ -tuple

$$(B_1, B_2, \dots, B_{K-1})$$

of its *right borders*, where border  $B_k$  is the index of the last item that belongs to group  $G_k$ . Notice that it is not necessary to specify  $B_K$ , the index of the last item of the last group, because its value will be  $N$  for any segmentation.

Almost all the dynamic programming algorithms for multiple channels assume that the items  $d_1, d_2, \dots, d_N$  are indexed by non-increasing  $\frac{p_i}{z_i}$  ratios, that is  $\frac{p_1}{z_1} \geq \frac{p_2}{z_2} \geq \dots \geq \frac{p_N}{z_N}$ . Observe that for unit lengths this means that the items are sorted by non-increasing popularities. Let  $SOL_{k,n}$  denote a segmentation for grouping items  $d_1, \dots, d_n$  into  $k$  groups and let  $sol_{k,n}$  be its corresponding cost, for any  $k \leq K$  and  $n \leq N$ . Moreover, let  $C_{i,j}$  denote the cost of assigning to a single channel the consecutive items  $d_i, \dots, d_j$ :

$$C_{i,j} = \sum_{h=i}^j t_h p_h = \sum_{h=i}^j \left( \frac{1}{2} \sum_{h=i}^j z_h \right) p_h = \frac{1}{2} \left( \sum_{h=i}^j z_h \right) \left( \sum_{h=i}^j p_h \right) \quad (3)$$

For unit lengths, the above formula simplifies as  $C_{i,j} = \frac{1}{2}(j - i + 1) \sum_{h=i}^j p_h$ . Note that, once the items are sorted, all the  $C_{i,j}$ 's can be found in  $O(N)$  time by means of prefix-sum computations [21].

The five main algorithms for solving the problem are now briefly surveyed. The first three of them, called *DP*, *Dichotomic*, and *Dlinear*, assume items sorted by non-increasing  $\frac{p_i}{z_i}$ 's (and thus they search for segmentations) and work for an arbitrary number of channels. Whereas, the other two, called *Knapsack* and *SRR*, do not assume sorted items and work for two channels and one channel, respectively. The first four algorithms are off-line and employ dynamic programming, while the last algorithm is on-line and does not use dynamic programming.

### 1.1. The DP Algorithm

The DP algorithm is a dynamic programming implementation of the following recurrence, where  $k$  varies from 1 to  $K$  and, for each fixed  $k$ ,  $n$  varies from 1 to  $N$ :

$$sol_{k,n} = \begin{cases} C_{1,n} & \text{if } k = 1 \\ \min_{1 \leq \ell \leq n-1} \{sol_{k-1,\ell} + C_{\ell+1,n}\} & \text{if } k > 1 \end{cases} \quad (4)$$

For any value of  $k$  and  $n$ , the DP algorithm selects the best segmentation obtained by considering the  $n - 1$  segmentations already computed for the first  $k - 1$  channels and for the first  $\ell$  items, and by combining each of them with the cost of assigning the last  $n - \ell$  items to the single  $k$ -th channel. In details, consider the  $K \times N$  matrix  $M$  with  $M_{k,n} = sol_{k,n}$ . The entries of  $M$  are computed row by row applying Recurrence 4. Clearly,  $M_{K,N}$  contains the cost of a solution for the original problem. In order to actually construct the corresponding segmentation, a second matrix  $F$  is employed to keep track of the final borders of segmentations corresponding to entries of  $M$ . In Recurrence 4, the value of  $\ell$  which minimizes the right-hand-side is the *final border* for the solution  $SOL_{k,n}$  and is stored in  $F_{k,n}$ . Hence, the segmentation is given by  $SOL_{K,N} = (B_1, B_2, \dots, B_{K-1})$  where, starting from  $B_K = N$ , the value of  $B_k$  is equal to  $F_{k+1, B_{k+1}}$ , for  $k = K - 1, \dots, 1$ . The DP algorithm requires  $O(N^2 K)$  time. It finds an optimal solution in the case of unit lengths and a sub-optimal one in the case of non-unit lengths [22].

### 1.2. The Dichotomic Algorithm

To improve on the time complexity of the DP algorithm, the Dichotomic algorithm has been devised. Let  $B_h^n$  denote the  $h$ -th border of  $SOL_{k,n}$ , with  $k > h \geq 1$ . Assume that  $SOL_{k-1,n}$  has been found for every  $1 \leq n \leq N$ . If  $SOL_{k,l}$  and  $SOL_{k,r}$  have been found for some  $1 \leq l \leq r \leq N$ , then one knows that  $B_{k-1}^c$  is between  $B_{k-1}^l$  and  $B_{k-1}^r$ , for any  $l \leq c \leq r$ . Thus, choosing  $c$  as the middle point between  $l$  and  $r$ , Recurrence 4 can be rewritten as:

$$sol_{k, \lceil \frac{l+r}{2} \rceil} = \min_{B_{k-1}^l \leq \ell \leq B_{k-1}^r} \{sol_{k-1,\ell} + C_{\ell+1, \lceil \frac{l+r}{2} \rceil}\} \quad (5)$$

where  $B_{k-1}^l$  and  $B_{k-1}^r$  are, respectively, the final borders of  $SOL_{k,l}$  and  $SOL_{k,r}$ .

Such a recurrence is iteratively solved within three nested loops which vary, respectively, in the ranges  $1 \leq k \leq K$ ,  $1 \leq t \leq \lceil \log N \rceil$ , and  $1 \leq i \leq 2^{t-1}$ , and where the indices  $l$ ,  $r$ , and  $c$  are set as follows:  $l = \lceil \frac{i-1}{2^{t-1}}(N+1) \rceil$ ,  $r = \lceil \frac{i}{2^{t-1}}(N+1) \rceil$ , and  $c = \lceil \frac{l+r}{2} \rceil = \lceil \frac{2i-1}{2^t}(N+1) \rceil$ . In details, the Dichotomic algorithm is shown in Figure 1.

```

Input:      N items sorted by non-increasing  $\frac{p_i}{z_i}$  ratios, and K groups;
Initialize: for i from 1 to N do
             for k from 1 to K do
                 if k = 1 then  $M_{k,i} \leftarrow C_{k,i}$  else  $M_{k,i} \leftarrow \infty$ ;
Loop 1:     for k from 2 to K do
              $F_{k,0} \leftarrow F_{k,1} \leftarrow 1$ ;  $F_{k,N+1} \leftarrow N$ ;
Loop 2:     for t from 1 to  $\lceil \log N \rceil$  do
Loop 3:     for i from 1 to  $2^{t-1}$  do
              $c \leftarrow \lceil \frac{2i-1}{2^t}(N+1) \rceil$ ;
              $l \leftarrow \lceil \frac{i-1}{2^{t-1}}(N+1) \rceil$ ;
              $r \leftarrow \lceil \frac{i}{2^{t-1}}(N+1) \rceil$ ;
             if  $M_{k,c} = \infty$  then
Loop 4:     for  $\ell$  from  $F_{k,l}$  to  $F_{k,r}$  do
                 if  $M_{k-1,\ell} + C_{\ell+1,c} < M_{k,c}$  then
                      $M_{k,c} \leftarrow M_{k-1,\ell} + C_{\ell+1,c}$ ;
                      $F_{k,c} \leftarrow \ell$ ;

```

**Figure 1.** The Dichotomic algorithm.

It uses the two matrices  $M$  and  $F$ , whose entries are again filled up row by row (Loop 1). A generic row  $k$  is filled in stages (Loop 2). Each stage corresponds to a particular value of the variable  $t$  (Loop 3). The variable  $c$  corresponds to the index of the entry which is currently being filled in stage  $t$ . The variables  $l$  (left) and  $r$  (right) correspond to the indices of the entries nearest to  $c$  which have been already filled, with  $l < c < r$ . If no entry before  $c$  has been already filled, then  $l = 1$ , and therefore the final border  $F_{k,1}$  is initialized to 1. If no entry after  $c$  has been filled, then  $r = N$ , and thus the final border  $F_{k,N+1}$  is initialized to  $N$ . To compute the entry  $c$ , the variable  $\ell$  takes all values between  $F_{k,l}$  and  $F_{k,r}$ . The index  $\ell$  which minimizes the recurrence in Loop 4 is assigned to  $F_{k,c}$ , while the corresponding minimum value is assigned to  $M_{k,c}$ .

The Dichotomic algorithm lowers the time complexity of the DP algorithm to  $O(NK \log N)$ . As for the DP algorithm, the Dichotomic algorithm also finds optimal and sub-optimal solutions for unit and non-unit lengths, respectively [5].

### 1.3. The Dlinear Algorithm

Fixed  $k$  and  $n$ , the Dlinear algorithm selects the feasible segmentations that satisfy the following Recurrence:

$$sol_{k,n} = \begin{cases} C_{1,n} & \text{if } k = 1 \\ sol_{k-1,m} + C_{m+1,n} & \text{if } k > 1 \end{cases} \quad (6)$$

where

$$m = \min_{B_k^{n-1} \leq \ell \leq n-1} \{ \ell : sol_{k-1,\ell} + C_{\ell+1,n} < sol_{k-1,\ell+1} + C_{\ell+2,n} \}.$$

```

Input:       $N$  items sorted by non-increasing  $\frac{p_i}{z_i}$  ratios, and  $K$  groups;
Initialize:  for  $n$  from 1 to  $N$  do
                 $M_{1,n} \leftarrow C_{1,n}$ ;
Loop 1:      for  $k$  from 2 to  $K$  do
                 $F_{k,k} \leftarrow k - 1$ ;
                 $M_{k,k} \leftarrow M_{k-1,k-1} + C_{k,k}$ ;
Loop 2:      for  $n$  from  $k + 1$  to  $N$  do
                 $\ell \leftarrow F_{k,n-1}$ ;
                 $m \leftarrow \ell$ ;
                 $M_{k,n} \leftarrow M_{k-1,\ell} + C_{\ell+1,n}$ ;
                 $incr \leftarrow \text{false}$ ;
Loop 3:      while  $\ell \leq n - 2$  and  $\neg incr$  do
                 $temp \leftarrow M_{k-1,\ell+1} + C_{\ell+2,n}$ ;
                if  $M_{k,n} \geq temp$  then
                     $M_{k,n} \leftarrow temp$ ;
                     $\ell \leftarrow \ell + 1$ ;
                else
                     $incr \leftarrow \text{true}$ ;
                 $m \leftarrow \ell$ ;
                 $F_{k,n} \leftarrow m$ 

```

**Figure 2.** The Dlinear algorithm.

In practice, Dlinear adapts Recurrence 4 by exploiting the property that, if  $SOL_{k,n-1}$  is known, then one knows that  $B_k^n$  is no smaller than  $B_k^{n-1}$ , and by stopping the trials as soon as the cost  $sol_{k-1,\ell} + C_{\ell+1,n}$  of the solution starts to increase.

The Dlinear algorithm is shown in Figure 2. As before, matrices  $M$  and  $F$  are used, which are filled row by row. Note that in Loop 1 the leftmost  $k - 1$  entries in row  $k$  of both  $M$  and  $F$  are meaningless, since at least one item has to be assigned to each channel. The value of  $m$  in Recurrence 6 that gives  $M_{k,n}$  is computed iteratively in Loop 3 and stored in  $F_{k,n}$ .

The overall time complexity of the Dlinear algorithm is  $O(N(K + \log N))$ . Thus the Dlinear algorithm is even faster than the Dichotomic one, but the solutions it provides are always sub-optimal, both in the unit and non-unit length case [4].

#### 1.4. The Knapsack Algorithm

The Knapsack algorithm solves the problem when there are exactly 2 channels. In such a case, the problem is to find a partition  $G_1$  and  $G_2$  such that  $\frac{1}{2}(Z_1P_1 + Z_2P_2)$  is minimized, where  $P_k$  and  $Z_k$  denote the sum of the popularities and of the lengths, respectively, of items in  $G_k$ , for  $k = 1$  and 2. Clearly,  $P_1 + P_2 = 1$  and  $Z_1 + Z_2 = Z$ . Without loss of generality,  $Z_1 \leq Z_2$  can be assumed, and hence there are only  $\lfloor Z/2 \rfloor$  possible values for  $Z_1$ .

Observe that  $Z_1P_1 + Z_2P_2 = Z_1P_1 + Z_2(1 - P_1) = P_1(Z_1 - Z_2) + Z_2$ . When  $Z_1$  is fixed, also  $Z_2 = Z - Z_1$  is fixed, and noting that  $Z_1 - Z_2 \leq 0$ , minimizing  $Z_1P_1 + Z_2P_2$  is equivalent to maximizing  $P_1$ . Therefore, the problem reduces to a particular *Knapsack problem* [14] of capacity  $Z_1$ , where each item  $d_i$  is characterized by a *profit*  $p_i$  and a *weight*  $z_i$ . Specifically, the Knapsack problem consists in finding a subset  $S$

of  $\{d_1, d_2, \dots, d_N\}$  subject to the constraint  $\sum_{d_k \in S} z_k = Z_1$  so as to maximize the objective function  $\sum_{d_k \in S} p_k$ .

To apply dynamic programming, consider two  $(N + 1) \times (\lfloor Z/2 \rfloor + 1)$  matrices  $M$  and  $X$ . The entry  $M_{i,j}$ , with  $0 \leq i \leq N$  and  $0 \leq j \leq \lfloor Z/2 \rfloor$ , stores the value of the objective function for the above Knapsack problem with items  $\{d_1, \dots, d_i\}$  and capacity  $j$ . Formally,  $M_{i,j} = \max \sum_{d_k \in S} p_k$  such that  $\sum_{d_k \in S} z_k = j$ , where  $S \subseteq \{d_1, \dots, d_i\}$ . By definition,  $M_{i,j} = -\infty$  if the capacity  $j$  cannot be completely filled by any  $S$ . The boolean entry  $X_{i,j}$  records whether the item  $d_i$  has been selected or not in the solution of the Knapsack problem with items  $\{d_1, \dots, d_i\}$  and capacity  $j$ , with  $0 \leq i \leq N$  and  $0 \leq j \leq \lfloor Z/2 \rfloor$ .

The dynamic programming algorithm starts by initializing the first row of the matrices in such a way that  $M_{0,0} = 0$ ,  $M_{0,j} = -\infty$  for  $1 \leq j \leq \lfloor Z/2 \rfloor$ , and  $X_{0,j} = \text{false}$  for  $0 \leq j \leq \lfloor Z/2 \rfloor$ . Then, for  $i = 1, 2, \dots, N$  and  $j = 0, 1, \dots, \lfloor Z/2 \rfloor$ ,  $M_{i,j}$  and  $X_{i,j}$  are filled by using the following relations:

$$M_{i,j} = \begin{cases} M_{i-1,j} & \text{if } j < z_i \\ \max\{M_{i-1,j}, M_{i-1,j-z_i} + p_i\} & \text{if } j \geq z_i \end{cases} \quad (7)$$

$$X_{i,j} = \begin{cases} \text{true} & \text{if } M_{i,j} = M_{i-1,j-z_i} + p_i \neq -\infty \\ \text{false} & \text{otherwise} \end{cases}$$

Note that it is possible that for certain values of  $j$ , with  $0 \leq j \leq \lfloor Z/2 \rfloor$ , there is no solution for items  $\{d_1, \dots, d_i\}$  such that the total sum of weights is exactly  $j$ . In such cases, according to the definition, Recurrence 7 gives  $M_{i,j} = -\infty$ . In contrast, if there is a solution for items  $\{d_1, \dots, d_i\}$  such that the total sum of weights is exactly  $j$ , then  $M_{i,j} \neq -\infty$  and  $M_{i,j}$  gives the optimal value of the objective function.

Consider the last row of  $M$ . Any entry  $M_{N,j} \neq -\infty$  gives the optimal  $P_1$  for the 2-channel data allocation problem with items  $\{d_1, \dots, d_N\}$  and  $Z_1 = j$ . Therefore, the entry, say  $M_{N,\bar{j}}$ , which minimizes  $\frac{1}{2} (\bar{j} M_{N,\bar{j}} + (Z - \bar{j})(1 - M_{N,\bar{j}}))$  gives the optimal AED for the original problem. Once  $M_{N,\bar{j}}$  has been found, it is easy to list out the items which have been picked up in the optimal solution, by tracing back the solution path. Specifically, if  $X_{N,\bar{j}} = \text{true}$ , then item  $d_N$  is selected and the entry  $X_{N-1,\bar{j}-z_N}$  is examined next; if  $X_{N,\bar{j}} = \text{false}$ , then item  $d_N$  is not selected and the entry  $X_{N-1,\bar{j}}$  is examined next. Such a procedure is repeated backwards until the row 0 of  $X$  is reached. The selected items are assigned to group  $G_1$ , while the remaining items are assigned to group  $G_2$ .

The Knapsack algorithm always finds an optimal solution for two channels and non-unit lengths and its overall time complexity is  $O(NZ)$ , which is pseudo-polynomial [5]. The algorithm is effective when the items have small length. For instance, if each item length is bounded by a constant, then  $Z = O(N)$  and the overall time becomes  $O(N^2)$ .

### 1.5. The SRR Algorithm

When there is only one channel, the DP, Dichotomic, and Dlinear algorithms provide a trivial flat schedule with period  $Z$ . In such a case, each  $t_i$  is equal to  $\frac{Z}{2}$  and hence also the AED is equal to  $\frac{Z}{2}$ , regardless of the item popularities. To overcome this drawback,



a schedule is needed where the spacing between two consecutive transmissions of one item is not the same for all items, but depends on both the popularity and the length of such an item.

It has been shown in [19] that, in an optimal schedule, replicas of any item  $d_i$  should be equally spaced with spacing

$$s_i = \left( \sum_{h=1}^N \sqrt{p_h z_h} \right) \sqrt{\frac{z_i}{p_i}} \quad (8)$$

In this way, the expected delay for item  $d_i$  becomes half of its spacing and thus, substituting  $t_i = \frac{s_i}{2}$  in Equation 1, the average expected delay becomes

$$\text{AED} = \frac{1}{2} \left( \sum_{i=1}^N \sqrt{p_i z_i} \right)^2 \quad (9)$$

The AED value given in Equation 9 represents a lower bound which in general is not achievable because the replicas cannot always be kept equally spaced. The SRR algorithm is an on-line heuristic which tries to keep the replicas as equally spaced as possible. For this purpose, it determines the item to be transmitted next by using the decision rule  $\frac{s_i p_i}{z_i} = \text{constant}$ , based on Equation 8. Let  $T$  denote the current time, let  $R_i$  be the time at which the last replica of  $d_i$  has been transmitted (initialized to  $-1$ ), and let  $G_i = (T - R_i)^2 \frac{p_i}{z_i}$ , where  $T - R_i$  is the spacing for item  $d_i$  if  $d_i$  would be transmitted again at time  $T$ . At each instant of time  $T$ , the SRR algorithm evaluates the decision rule  $G_i$  for all items  $d_i$ ,  $1 \leq i \leq N$ , selects for transmission at time  $T$  that item  $d_h$  with maximum  $G_h$ , and finally updates  $R_h = T$  and  $T = T + z_h$ .

The SRR algorithm takes  $O(N)$  time to select the item to be transmitted. Such a time can be reduced to  $O(M)$  by partitioning the items into  $M$  buckets according to their  $G$ 's values [19].

## 2. Bernoulli Channel Error Model

In this section, unrecoverable channel transmission errors modeled by a geometric distribution are taken into account. Under such an error model, each packet transmission over every channel has the same probability  $q$  to fail and  $1 - q$  to succeed, and each transmission error is independent from the others, with  $0 \leq q \leq 1$ . Since the environment is asymmetric, a client cannot ask the server to immediately retransmit an item  $d_i$  which has been received on channel  $k$  with an unrecoverable error. Indeed, the client has to discard the item and then has to wait for a whole period  $Z_k$ , until the next transmission of  $d_i$  scheduled by the server. Even the next item transmission could be corrupted, and in such a case an additional delay of  $Z_k$  has to be waited. Therefore, the expected delay  $t_i$  has to take into account the extra waiting time due to a possible sequence of independent unrecoverable errors.

### 2.1. Unit Length Items

Assume that the items have unit lengths, i.e.,  $z_i = 1$  for  $1 \leq i \leq N$ . Recall that in such a case the period of channel  $k$  is  $N_k$ . If a client wants to receive item  $d_i$ , which is transmitted on channel  $k$ , and the first transmission he can hear of  $d_i$  is error-free, then the client waits on the average  $\frac{N_k}{2}$  time units with probability  $1 - q$ . Instead, if the first transmission of  $d_i$  is erroneous, but the second one is error-free, then the client experiences an average delay of  $\frac{N_k}{2} + N_k$  time units with probability  $q(1 - q)$ . Generalizing, if there are  $h$  bad transmissions of  $d_i$  followed by a good one, the client average delay for receiving item  $d_i$  becomes  $\frac{N_k}{2} + hN_k$  time units with probability  $q^h(1 - q)$ . Thus, summing up over all  $h$ , the expected delay  $t_i$  is equal to

$$\sum_{h=0}^{\infty} \left( \frac{N_k}{2} + hN_k \right) q^h (1 - q) = \frac{N_k}{2} + N_k \frac{q}{1 - q}$$

because  $\sum_{h=0}^{\infty} q^h = \frac{1}{1 - q}$  and  $\sum_{h=0}^{\infty} hq^h = \frac{q}{(1 - q)^2}$ . Therefore, one can set the expected delay as

$$t_i = \frac{N_k}{2} \frac{1 + q}{1 - q} \quad (10)$$

By the above setting, the objective function to be minimized becomes

$$\text{AED} = \sum_{i=1}^N t_i p_i = \frac{1}{2} \frac{1 + q}{1 - q} \sum_{k=1}^K N_k P_k \quad (11)$$

Therefore, for items with unit lengths, the data allocation problem can be optimally solved in polynomial time. This derives from Lemmas 1 and 2 of [5] which prove optimality in the particular case of error-free channels, that is, when  $q = 0$ . Indeed, when  $q > 0$ , similar proofs hold once the cost  $C_{i,j}$  of assigning consecutive items  $d_i, \dots, d_j$  to the same channel is defined as  $C_{i,j} = \frac{j-i+1}{2} \frac{1+q}{1-q} \sum_{h=i}^j p_h$ . In words, Lemmas 1 and 2 of [5] show that, whenever the items  $d_1, d_2, \dots, d_N$  are sorted by non-increasing popularities, there always exists an optimal solution which is a segmentation and which can be found by the Dichotomic algorithm.

### 2.2. Non-Unit Length Items

Consider now items with non-unit lengths and recall that  $Z_k$  is the period of channel  $k$ . In order to receive an item  $d_i$  of length  $z_i$  over channel  $k$ , a client has to listen for  $z_i$  consecutive error-free packet transmissions, which happens with probability  $(1 - q)^{z_i}$ . Hence, the failure probability for item  $d_i$  on channel  $k$  is  $Q_{z_i} = 1 - (1 - q)^{z_i}$ .

In the case that the first transmission of  $d_i$  heard by the client is error-free, the client has to wait on the average  $\frac{Z_k}{2}$  time units with probability  $1 - Q_{z_i}$ . Instead, the client waits on the average for  $\frac{Z_k}{2} + Z_k$  time units with probability  $Q_{z_i}(1 - Q_{z_i})$  in the case that the first transmission of  $d_i$  is erroneous and the second one is error-free. In general,  $h$  bad transmissions of  $d_i$  followed by a good one lead to a delay of  $\frac{Z_k}{2} + hZ_k$  time units with

probability  $Q_{z_i}^h (1 - Q_{z_i})$ . Therefore, summing up over all  $h$  as seen in the unit length case, the expected delay becomes

$$t_i = \frac{Z_k}{2} \frac{1 + Q_{z_i}}{1 - Q_{z_i}} \quad (12)$$

Thus, the average expected delay to be minimized is

$$\text{AED} = \frac{1}{2} \sum_{k=1}^K \left( Z_k \sum_{d_i \in G_k} \frac{1 + Q_{z_i}}{1 - Q_{z_i}} p_i \right) \quad (13)$$

Recalling that the items are indexed by non-increasing  $\frac{p_i}{z_i}$  ratios, the new recurrences for the Dichotomic and Dlinear algorithms are derived from Recurrences 5 and 6, respectively, once each  $C_{i,j}$  is defined as  $C_{i,j} = \frac{1}{2} \left( \sum_{h=i}^j z_h \right) \left( \sum_{h=i}^j \frac{1 + Q_{z_h}}{1 - Q_{z_h}} p_h \right)$ . All the  $C_{i,j}$ 's can be computed in  $O(N)$  time via prefix-sums, once  $O(H)$  time is spent for computing all the  $Q_{z_h}$ 's, where  $H = \min\{N \log z, z\}$ . Therefore, the time complexities of the Dichotomic and Dlinear algorithms become, respectively,  $O(NK \log N + H)$  and  $O(N(K + \log N) + H)$ . Note that in such a case optimality is not guaranteed since the problem is computationally intractable already for error-free channels. However, when there are only two channels, an optimal solution can be found in  $O(NZ)$  time applying the Knapsack algorithm, simply replacing each popularity  $p_i$  with  $p'_i = \frac{1 + Q_{z_i}}{1 - Q_{z_i}} p_i$  in Recurrence 7, and then finally selecting the entry  $M_{N,\bar{j}}$  which minimizes  $\frac{1}{2} \left( \bar{j} M_{N,\bar{j}} + (Z - \bar{j})(P' - M_{N,\bar{j}}) \right)$ , where  $P' = \sum_{i=1}^N p'_i$ .

When there is only one channel, it has been shown in [19] that, in an optimal schedule, replicas of any item  $d_i$  should be equally spaced with spacing

$$s_i = \left( \sum_{h=1}^N \sqrt{p_h z_h \frac{1 + Q_{z_h}}{1 - Q_{z_h}}} \right) \sqrt{\frac{z_i}{p_i} \frac{1 - Q_{z_i}}{1 + Q_{z_i}}} \quad (14)$$

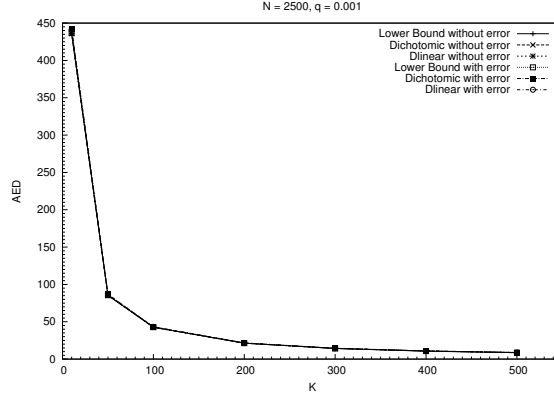
Thus, substituting  $t_i = \frac{s_i}{2}$  in Equation 1, the average expected delay becomes

$$\text{AED} = \frac{1}{2} \left( \sum_{i=1}^N \sqrt{p_i z_i \frac{1 + Q_{z_i}}{1 - Q_{z_i}}} \right)^2 \quad (15)$$

Therefore, the SRR algorithm can be applied once the decision rule  $G_i$  is modified as  $G_i = (T - R_i)^2 \frac{p_i}{z_i} \frac{1 + Q_{z_i}}{1 - Q_{z_i}}$ .

### 2.3. Performance Evaluation

In this subsection, the behavior of the Dichotomic, Dlinear, and SRR heuristics is evaluated in the case of Bernoulli channel error model. The above algorithms have been experimentally tested on benchmarks where the item popularities follow a Zipf distribution. Specifically, given the number  $N$  of items and a real number  $0 \leq \theta \leq 1$ , the item popularities are defined as



**Figure 3.** Results for 2500 items of non-unit lengths, when  $\theta = 0.8$  and the  $K$  channels have failure probability  $q = 0.001$ .

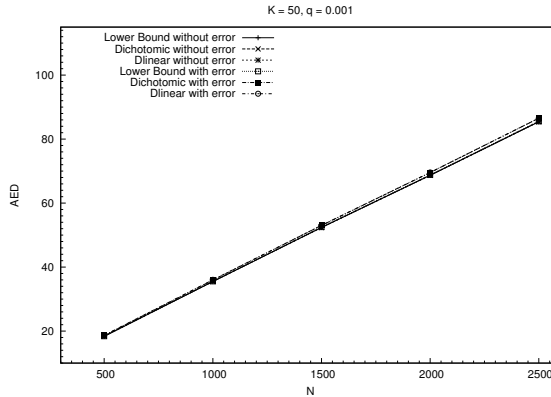
$$p_i = \frac{(1/i)^\theta}{\sum_{h=1}^N (1/h)^\theta} \quad 1 \leq i \leq N$$

In the above formula,  $\theta$  is the *skew* parameter. In particular,  $\theta = 0$  stands for a uniform distribution with  $p_i = \frac{1}{N}$ , while a higher  $\theta$  implies a higher skew, namely the difference among the  $p_i$  values becomes larger.

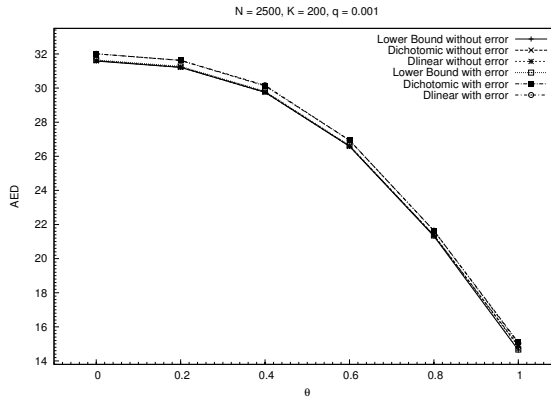
Consider first some experiments for multiple channels reported from [7], where either the skew parameter  $\theta$  is set to 0.8 as suggested in [22],  $N = 2500$ , and  $10 \leq K \leq 500$ , or  $\theta = 0.8$ ,  $K = 50$ , and  $500 \leq N \leq 2500$ , or  $0 \leq \theta \leq 1$ ,  $N = 2500$ , and  $K = 200$ . The item lengths  $z_i$  are integers randomly generated according to a uniform distribution in the range  $1 \leq z_i \leq 10$ , for  $1 \leq i \leq N$ . The channel failure probabilities can assume the values 0.001 and 0.01.

Moreover, since the data allocation problem is computationally intractable when items have non-unit lengths, lower bounds for a non-unit length instance are derived by transforming it into a unit length instance as follows. Each item  $d_i$  of popularity  $p_i$  and length  $z_i$  is decomposed into  $z_i$  items of popularity  $\frac{p_i}{z_i}$  and length 1. Since more freedom has been introduced, it is clear that the optimal AED for the so transformed problem is a lower bound on the AED of the original problem. Since the transformed problem has unit lengths, the optimal AED can be obtained by running the polynomial time Dichotomic algorithm both when all the channels are error-free or have the same failure probability.

Figures 2.3-5 show the experimental results for the Dichotomic and Dlinear algorithms in the case that there are multiple channels, the items have non-unit lengths, and the failure probability  $q$  is 0.001. One can note that the two above mentioned lower bounds as well as the solutions provided by both algorithms almost coincide. Instead, Figures 6-8 show the experimental results when the failure probability  $q$  is 0.01. Referring to Figures 6 and 7, where  $\theta = 0.8$ , the AED of the transformed unit length instance in the presence of errors is  $\frac{1+q}{1-q} = 1.02$  times the AED of the same transformed instance without errors. One can also note that, since the average item length is 5, the AED of the original instance in the presence of errors should be about  $\frac{1+Q}{1-Q} = 1.10$  times the AED of the same original instance in the absence of errors, where  $Q = 1 - (1 - 0.01)^5 = 0.05$ .



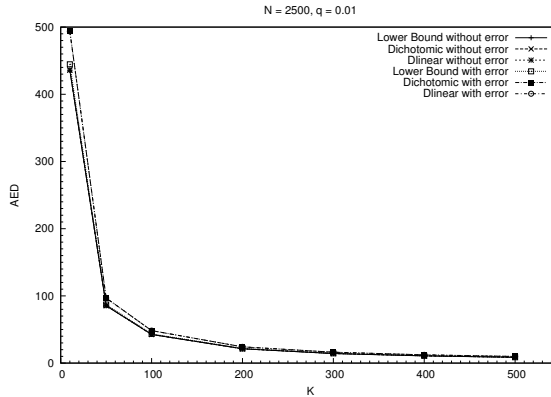
**Figure 4.** Results for  $N$  items of non-unit lengths, when  $\theta = 0.8$  and the 50 channels have failure probability  $q = 0.001$ .



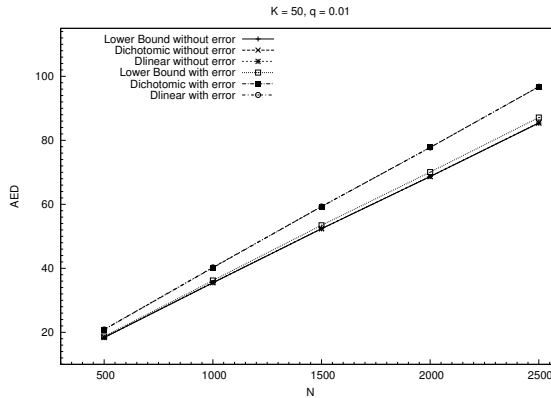
**Figure 5.** Results for 2500 items of non-unit lengths, when  $0 \leq \theta \leq 1$  and the 200 channels have failure probability  $q = 0.001$ .

This can be easily checked in Figure 6, e.g., for  $K = 10$ , where the ratio between the two AEDs is about  $\frac{500}{450} = 1.11$ . Referring to Figure 8, where  $\theta$  varies, one notes that the ratio between such AEDs is almost 1.12 for every value of  $\theta$ , confirming the results of Figures 6 and 7.

Consider now some simulation experiments for a single channel, which are reported from [19]. In the experiments,  $N = 1000$ ,  $0 \leq \theta \leq 1$ , and each  $z_i$  is an integer randomly generated according to a uniform distribution in the range  $1 \leq z_i \leq 10$ , for  $1 \leq i \leq N$ . The channel failure probability  $q$  varies between 0 and 0.2. Figure 9 shows the behavior of the SRR algorithm compared with the analytical lower bound given in Equation 15. The experimental tests show that the AED values obtained by the SRR algorithm and by the lower bound differ up to 3% for small values of  $q$ , and up to 10% for larger values of  $q$ .



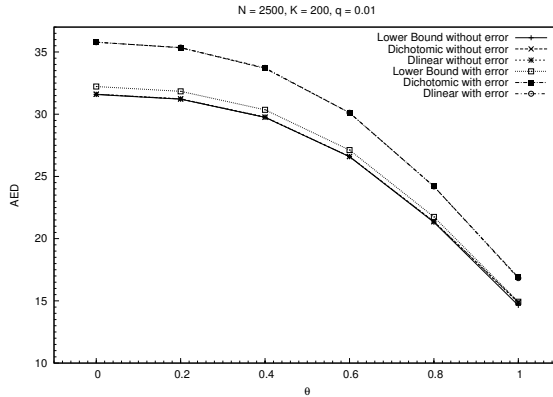
**Figure 6.** Results for 2500 items of non-unit lengths, when  $\theta = 0.8$  and the  $K$  channels have failure probability  $q = 0.01$ .



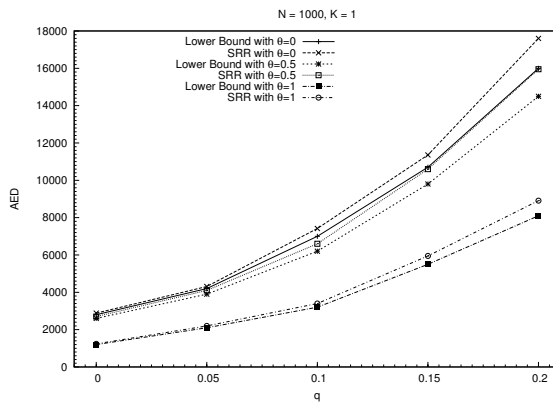
**Figure 7.** Results for  $N$  items of non-unit lengths, when  $\theta = 0.8$  and the 50 channels have failure probability  $q = 0.01$ .

### 3. Gilbert-Elliot Channel Error Model

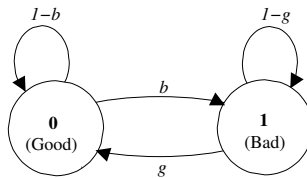
In this section, the channel error behavior is assumed to follow a simplified Gilbert-Elliot model, which is a two-state time-homogeneous discrete time Markov chain [20], as described below. At each time instant, a channel can be in one of two states. The state 0 denotes the *good* state, where the channel works properly and thus a packet is received with no errors. Instead, the state 1 denotes the *bad* state, where the channel is subject to failure and hence a packet is received with an unrecoverable error. Let  $X_0, X_1, X_2, \dots$  be the states of the channel at times  $0, 1, 2, \dots$ . The time between  $X_u$  and  $X_{u+1}$  corresponds to the length of one packet. The initial state  $X_0$  is selected randomly. As depicted in Figure 10, the probability of transition from the good state to the bad one is denoted by  $b$ , while that from the bad state to the good one is  $g$ . Hence,  $1 - b$  and  $1 - g$  are the probabilities of remaining in the same state, namely, in the good and bad state, respec-



**Figure 8.** Results for 2500 items of non-unit lengths, when  $0 \leq \theta \leq 1$  and the 200 channels have failure probability  $q = 0.01$ .



**Figure 9.** Results for 1000 items with non-unit lengths, when  $K = 1$ ,  $0 \leq \theta \leq 1$ , and  $0 \leq q \leq 0.2$ .



**Figure 10.** The Gilbert-Elliot channel error model.

tively. Formally,  $Prob[X_{u+1} = 0|X_u = 0] = 1 - b$ ,  $Prob[X_{u+1} = 0|X_u = 1] = g$ ,  $Prob[X_{u+1} = 1|X_u = 1] = 1 - g$ , and  $Prob[X_{u+1} = 1|X_u = 0] = b$ .

It is well known that the *steady-state* probability of being in the good state is  $P_G = \frac{g}{b+g}$ , while that of being in the bad state is  $P_B = \frac{b}{b+g}$ . This Markovian process has

mean  $\mu = P_B$ , variance  $\sigma^2 = \mu(1 - \mu) = \frac{bg}{(b+g)^2}$ , and autocorrelation function  $r(\nu) = P_B + (1 - P_B)(1 - b - g)^\nu$ , where  $b + g < 1$  is assumed. Since the system is memoryless, the state holding times are geometrically distributed. The mean state holding times for the good state and the bad state are, respectively,  $\frac{1}{b}$  and  $\frac{1}{g}$ . This means that the channel exhibits error bursts of consecutive ones whose mean length is  $\frac{1}{g}$ , separated by gaps of consecutive zeros whose mean length is  $\frac{1}{b}$ .

### 3.1. Unit Length Items

Assume that the items have unit lengths, i.e.,  $z_i = 1$  for  $1 \leq i \leq N$ . Recall that in such a case the period of channel  $k$  is  $N_k$ .

If a client waits for item  $d_i$  on channel  $k$ , and no error occurs in the first transmission of  $d_i$ , then the client waits on the average  $\frac{N_k}{2}$  time units with probability  $P_G = 1 - P_B$ . Instead, if an error occurs during the first transmission of  $d_i$  and there is no error in the second transmission, then the average delay experienced by the client is  $\frac{N_k}{2} + N_k$  time units with probability  $P_B(1 - r(N_k))$ . In general, when there are  $h$  erroneous transmissions of  $d_i$  followed by an error-free one, the client average delay is  $\frac{N_k}{2} + hN_k$  time units with probability  $P_B(r(N_k))^{h-1}(1 - r(N_k))$ . Thus, the expected delay is equal to

$$\frac{N_k}{2}P_G + P_B(1 - r(N_k)) \sum_{h=1}^{\infty} \left( \frac{N_k}{2} + hN_k \right) (r(N_k))^{h-1} =$$

$$\frac{N_k}{2}P_G + P_B \frac{N_k}{2} + P_B \frac{N_k}{1 - r(N_k)}$$

because  $\sum_{h=1}^{\infty} (r(N_k))^{h-1} = \frac{1}{1 - r(N_k)}$  and  $\sum_{h=1}^{\infty} h(r(N_k))^{h-1} = \frac{1}{(1 - r(N_k))^2}$ . Hence, the expected delay  $t_i$  and the objective function AED become, respectively:

$$t_i = \frac{N_k}{2} \left( 1 + \frac{2P_B}{1 - r(N_k)} \right) \quad (16)$$

$$\text{AED} = \frac{1}{2} \sum_{k=1}^K \left( N_k \left( 1 + \frac{2P_B}{1 - r(N_k)} \right) \sum_{d_i \in G_k} p_i \right) \quad (17)$$

It has been proved that, since all the items have unit length, there always exists an optimal solution which is a segmentation. Moreover, such a solution can be found in  $O(N^2K)$  time by the DP algorithm, whose new recurrence is derived from Recurrence 4 by setting  $C_{i,j} = \frac{j-i+1}{2} \left( 1 + \frac{2P_B}{1 - r(j-i+1)} \right) \sum_{h=i}^j p_h$ .

### 3.2. Non-Unit Length Items

This subsection deals with items of non-unit lengths. Recall that  $Z_k$  is the period of channel  $k$  and that a client has to listen for  $z_i$  consecutive error-free packet transmissions in order to receive the item  $d_i$  over channel  $k$ .



Consider now the first transmission of item  $d_i$  heard by a client. Let  $\hat{P}_B(s)$  denote the probability that in such a transmission the  $s$ -th packet is the first erroneous packet, where  $1 \leq s \leq z_i$ . Formally,

$$\hat{P}_B(s) = \begin{cases} P_B & \text{if } s = 1 \\ (1 - P_B)(1 - b)^{s-2}b & \text{if } 2 \leq s \leq z_i \end{cases}$$

Consider now two consecutive transmissions of item  $d_i$  heard by a client, the first of which is erroneous. Let  $\bar{P}_B(s, \sigma)$  denote the probability that, in the second transmission, the first erroneous packet is the  $s$ -th one given that in the previous transmission the first erroneous packet was the  $\sigma$ -th one. Thus, when  $s = 1$ ,  $\bar{P}_B(1, \sigma) = r(Z_k + 1 - \sigma)$ , whereas when  $2 \leq s \leq z_i$ :

$$\bar{P}_B(s, \sigma) = (1 - r(Z_k + 1 - \sigma))(1 - b)^{s-2}b$$

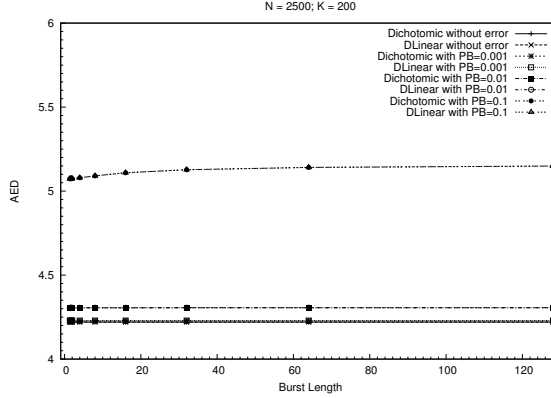
Finally, let  $\bar{P}_G(\sigma)$  denote the probability that a whole transmission of  $d_i$  is error-free given that in the previous transmission of  $d_i$  the first erroneous packet was the  $\sigma$ -th one:

$$\bar{P}_G(\sigma) = (1 - r(Z_k + 1 - \sigma))(1 - b)^{z_i-1}$$

Note that all the  $\hat{P}_B(s)$  and  $\bar{P}_B(s, \sigma)$ 's can be computed in pseudo-polynomial time, that is in a time polynomial in the parameters  $Z$  and  $z$ .

To evaluate the expected delay  $t_i$ , observe that if the first transmission of  $d_i$  heard by the client is error-free, the client has to wait on the average  $\frac{Z_k}{2}$  time units with probability  $(1 - P_B)(1 - b)^{z_i-1}$ . Instead, the client waits on the average for  $\frac{Z_k}{2} + Z_k$  time units with probability  $\sum_{s_0=1}^{z_i} \hat{P}_B(s_0)\bar{P}_G(s_0)$  in the case that the first transmission of  $d_i$  is erroneous and the second one is error-free. Moreover, two bad transmissions of  $d_i$  followed by a good one lead to a delay of  $\frac{Z_k}{2} + 2Z_k$  time units with probability  $\sum_{s_0=1}^{z_i} \left[ \hat{P}_B(s_0) \sum_{s_1=1}^{z_i} \bar{P}_B(s_1, s_0)\bar{P}_G(s_1) \right]$ . Thus, in general, the expected delay is  $t_i = \frac{Z_k}{2}(1 - P_B)(1 - b)^{z_i-1} + \sum_{h=1}^{\infty} \left[ \left( \frac{Z_k}{2} + hZ_k \right) \sum_{s_0=1}^{z_i} \left[ \hat{P}_B(s_0) \sum_{s_1=1}^{z_i} \left[ \bar{P}_B(s_1, s_0) \cdots \sum_{s_{h-1}=1}^{z_i} \left[ \bar{P}_B(s_{h-1}, s_{h-2})\bar{P}_G(s_{h-1}) \right] \cdots \right] \right] \right]$ .

Since finding a closed formula for  $t_i$  seems to be difficult, an approximation  $t_i^m$  of the expected delay can be computed by truncating the above series at the  $m$ -th term, for a given constant value  $m$ . Indeed, experimental tests show that the series converges already for small values of  $m$ , as it will be checked in Subsection 3.3. Thus, the average expected delay becomes  $\text{AED} = \sum_{i=1}^N t_i^m p_i$ . Recalling that the items are indexed by non-increasing  $\frac{p_i}{z_i}$  ratios, the Dichotomic and Dlinear algorithms can be applied once each  $C_{i,j}$  is computed as  $\sum_{h=i}^j t_h^m p_h$ . Fixed  $i$  and  $j$ , the time for computing  $C_{i,j}$  is derived as follows. Assuming a proper prefix-sum has been done in  $O(N)$  time as a preprocessing,  $Z_k = \sum_{h=i}^j z_h$  can be retrieved in  $O(1)$  time, while the computation of  $t_h^m$  requires  $O(z_h^m)$  time. Therefore, in the worst case, the computation of  $C_{i,j}$  takes  $O(Nz^m)$  time, and that of all the  $C_{i,j}$ 's costs  $O(N^3z^m)$  time, which is pseudo-polynomial. Hence, the time for computing the  $\hat{P}_B(s)$ 's,  $\bar{P}_B(s, \sigma)$ 's, and  $C_{i,j}$ 's leads to a pseudo-polynomial time complexity for both the Dichotomic and Dlinear algorithms.



**Figure 11.** The AED behaviour versus the mean error burst length, when  $\theta = 0.8$ ,  $N = 2500$ , and  $K = 200$ .

### 3.3. Performance Evaluation

This subsection presents some experimental tests, taken from [7], for the Dichotomic and DLinear heuristics in the case of the Gilbert-Elliot channel error model (experiments for the SRR heuristic are not available because it was studied only under the Bernoulli channel error model [19]).

In the experiments, the steady-state probability  $P_B$  of being in the bad state can assume the values 0.001, 0.01, and 0.1, while the mean error burst length  $\frac{1}{g}$  is fixed to 10. Note that  $b$  is derived as  $g \frac{P_B}{1-P_B}$  once  $P_B$  and  $\frac{1}{g}$  are fixed. However, the choice of  $\frac{1}{g}$  is not critical because the sensitivity of the AED to  $\frac{1}{g}$  is low, as depicted in Figure 11, for  $1 < \frac{1}{g} \leq 130$ . Note that the choice of such an upper bound for  $\frac{1}{g}$  is not restrictive because the probability of having a burst with length  $n$  is  $g(1-g)^{n-1}$ , which is negligible as  $n$  increases.

Since the algorithms take pseudo-polynomial time for items with non-unit lengths, a restricted set of experiments is performed. In the experiments, the number  $K$  of channels is set to 50, the number  $N$  of items varies between 500 and 2000, the item popularities follow a Zipf distribution with  $\theta = 0.8$ , and the item lengths  $z_i$  are integers randomly generated according to a uniform distribution in the range  $1 \leq z_i \leq 10$ , for  $1 \leq i \leq N$ . The expected delay of item  $d_i$  is evaluated by computing  $t_i^5$ , that is truncating at the fifth term the series giving  $t_i$ . Indeed, as shown in Table 1 for  $z_i = 10$ ,  $Z_k = 50$ ,  $\frac{1}{g} = 10$ , and  $P_B = 0.01$  and for  $z_i = 5$ ,  $Z_k = 50$ ,  $\frac{1}{g} = 10$ , and  $P_B = 0.1$ , at the fifth term the series giving  $t_i$  is already stabilized up to the fourth decimal digit.

Since the data allocation problem is computationally intractable when items have non-unit lengths, lower bounds for non-unit length instances are derived by transforming them into unit length instances, as explained in Subsection 2.3, and by running the DP algorithm. In particular, the AEDs giving the lower bounds are obtained from Equation 17.

Figure 12 shows the experimental results for non-unit lengths when  $P_B$ , which is identical for all channels, assumes the values 0.001, 0.01, and 0.1. In such a figure, lower bounds are shown for both error-free and error-prone channels. One notes that, for every value of  $P_B$ , the behavior of both the Dichotomic and DLinear algorithms is

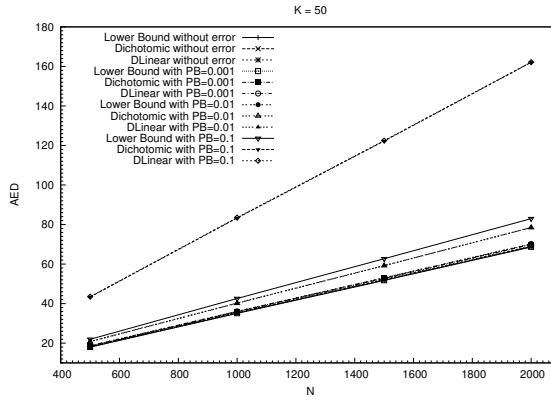
**Table 1.** Values of  $t_i^m$  when: (a)  $z_i = 10$ ,  $Z_k = 50$ ,  $\frac{1}{g} = 10$ , and  $P_B = 0.01$ ; and (b)  $z_i = 5$ ,  $Z_k = 50$ ,  $\frac{1}{g} = 10$ , and  $P_B = 0.16$ .

$m$	$t_i^m$
1	25.9150699
2	25.9382262
3	25.9388013
4	25.9388156
5	25.9388160
6	25.9388167

$m$	$t_i^m$
1	25.1989377
2	25.2537833
3	25.2689036
4	25.2730723
5	25.2745215
6	25.2745384

(a)

(b)



**Figure 12.** Results for  $N$  items with non-unit lengths, when  $\theta = 0.8$  and the 50 channels have the same steady-state probability  $P_B$ , which assumes the values 0.001, 0.01, and 0.1.

identical. When  $P_B = 0.001$ , both algorithms provide optimal solutions because their AEDs almost coincide with the lower bound for channels without errors. When  $P_B = 0.01$ , the AEDs of both the Dichotomic and Dlinear algorithms are 12% larger than the lower bound in presence of errors. In the last case, namely  $P_B = 0.1$ , the AEDs found by the algorithms are as large as twice those of the lower bound in presence of errors. However, such a value of  $P_B$  represents an extremal case which should not arise in practice (e.g. see [12]).

#### 4. Conclusions

This paper considered the problem of allocating data to multiple channels, assuming flat data scheduling per channel and the presence of unrecoverable channel transmission errors so as to minimize the average expected delay experienced by the clients. The behavior of some heuristics has been experimentally evaluated when modelling the channel error by means of the Bernoulli model as well as the simplified Gilbert-Elliot one.

Extensive simulations showed that such heuristics give good sub-optimal solutions when tested on benchmarks whose item popularities follow Zipf distributions. In particular, for small channel error probabilities, the average expected delay of the proposed solutions is almost the same as the optimal one found in the case of channels without errors. It is worth noting that, since the problem is computationally intractable (that is,  $NP$ -hard) for non-unit lengths and error-free channels, it remains intractable also in the presence of errors for non-unit lengths, while it can be polynomially solved for unit-lengths. As regard to the non-unit length case, an interesting open question is that of determining whether a closed formula for computing the item expected delays exists or not when the Gilbert-Elliott model is adopted. Moreover, an interesting extension of the problem to be investigated is that considering channels which do not have the same error probabilities. Some preliminary results on such an extension are reported in [7].

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